



ECE 344

MICROWAVE FUNDAMENTALS

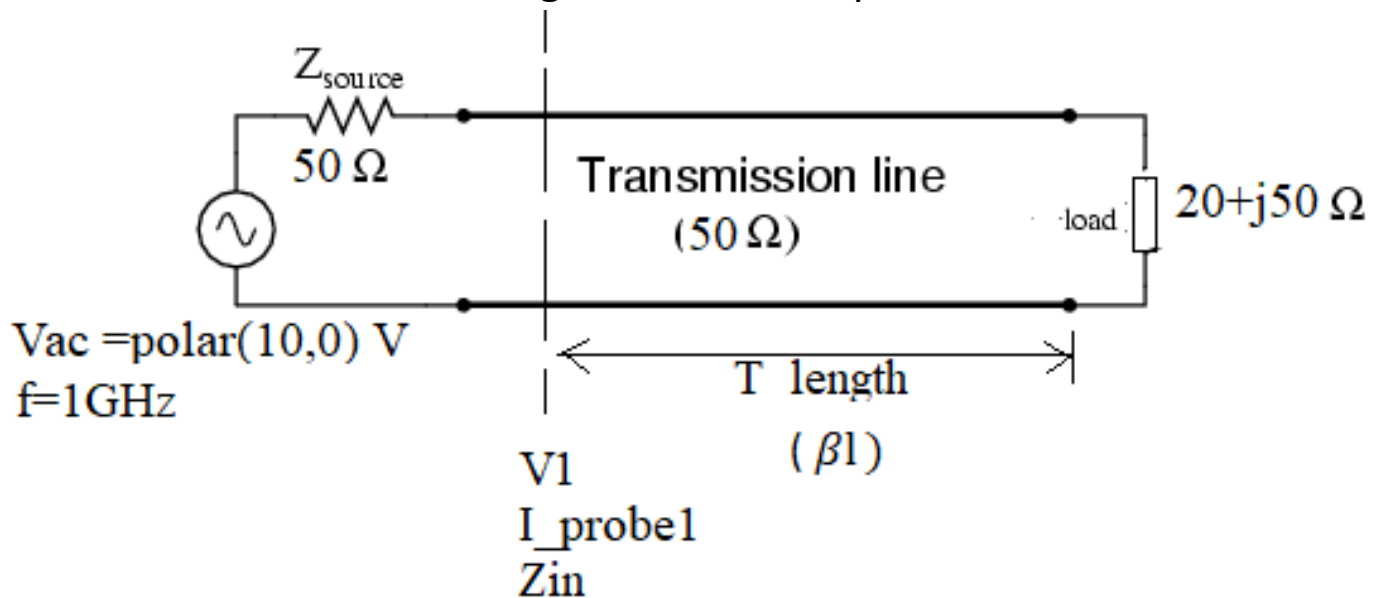
PART1-Lecture 5

Dr. Gehan Sami

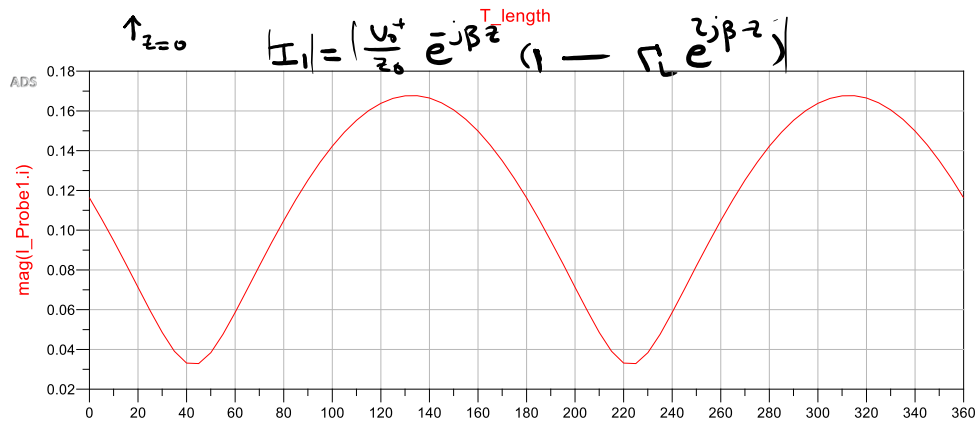
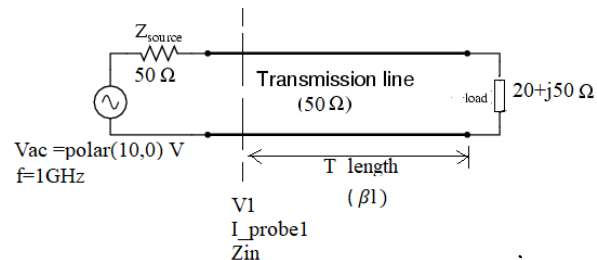
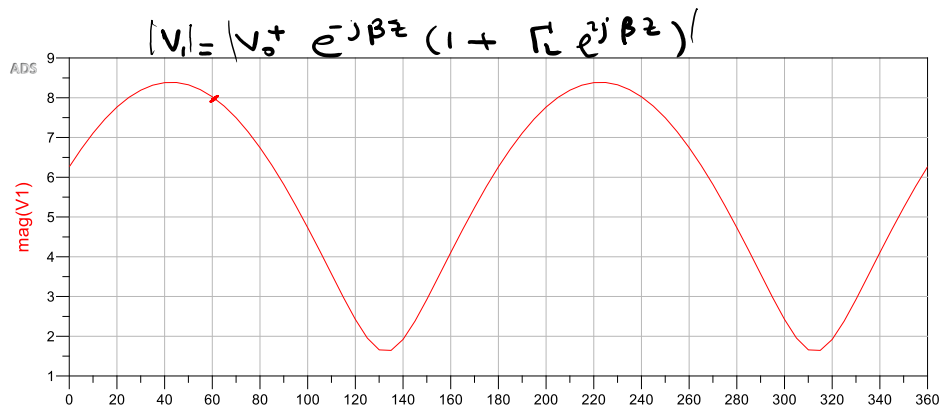
Example 1

USE ADS to:

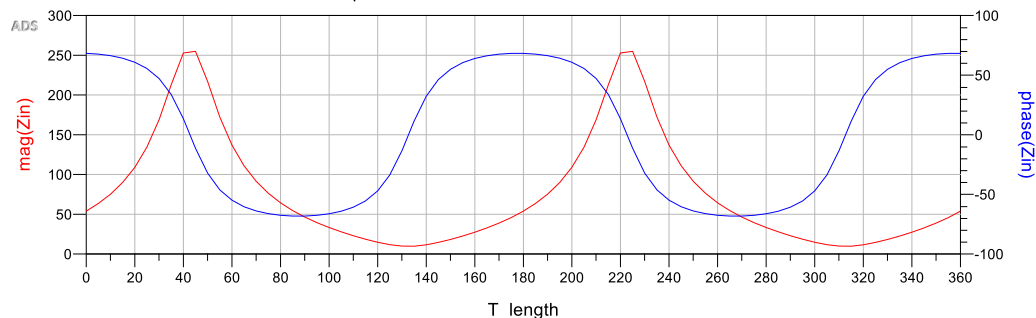
- draw magnitude voltage across the line $\beta l = 2\pi$ or $\ell = \lambda$
- draw magnitude current across the line
- draw impedance across the line
- observe $\text{mag}(V), \text{mag}(I), Z$ every $\ell = \lambda/2$**
- Compute magnitude of voltage, current at load.
- verify input impedance at load from voltage/current
Equals load impedance.
- find max voltage value and its position
- find min voltage value and its position



Terminated transmission line repeats its voltage mag., current mag. and impedance each $\lambda/2$



$Z_1 = V_1 / I_1$



@ $z=0$ $V_0^+ = 5 \text{ volt}$ $Z_0 = Z_0$

$V_1 = V_0^+ (1 + \Gamma)$, $\Gamma = \frac{Z_L - Z_0}{Z_L + Z_0}$

$V_1 = 5 (1 + 0.678 \angle 85.43)$

$V_1 = 6.26 \angle 32.67$

$I_1 = \frac{V_0^+}{Z_0} (1 - \Gamma) = 0.116 \angle -35.53$

$Z_1 = V_1 / I_1 = 53.9 \angle 68.2$
 $= Z_L$

$Z_L = 20 + j50 = 53.85 \angle 68.2$

$$\left. \begin{aligned} V(l + \lambda/2) &= -V(l) \\ I(l + \lambda/2) &= -I(l) \end{aligned} \right\} Z_{in}(l + \lambda/2) = Z_{in}(l)$$

$$V(l + \lambda) = V(l), \quad I(l + \lambda) = I(l)$$

$$\text{for } \beta l = 60^\circ \quad U_0^+ = 5 \quad \Gamma_L = 0.678 \angle 85.43$$

$$\text{Find } V(\beta l = 60), I(\beta l = 60), Z(\beta l = 60)$$

$$\begin{aligned} \text{Soln } V &= U_0^+ e^{j\beta l} (1 + \Gamma_L e^{-2j\beta l}) \\ &= 5 \angle 60 (1 + 0.678 \angle 85.43 \angle -120) \\ &= 5.56 + j5.79 = 8 \angle 46.13^\circ \text{ Volt} \end{aligned}$$

$$\begin{aligned} I &= \frac{U_0^+}{Z_0} e^{j\beta l} (1 - \Gamma_L e^{-2j\beta l}) \\ &= \frac{5}{50} \angle 60 (1 - 0.678 \angle 85.43 \angle -120) \\ &= -0.011 + j0.057 = 0.0586 \angle 101 \end{aligned}$$

$$Z = \frac{V}{I} = \frac{8 \angle 46.13}{0.0586 \angle 101} = 78.48 - j111.7 = 136.6 \angle -54.9^\circ$$

$$V_o^+ = 5 \text{ Volt} \quad \Gamma_L = 0.678 \quad \underline{85.43}$$

$$|V| = |V_o^+| \left| 1 + |\Gamma_L| e^{j\phi} e^{-2j\beta l} \right|$$

$$|V_{\max}| = |V_o^+| (1 + |\Gamma_L|) = 5 \times 1.678 = 8.4 \text{ Volt}$$

$$|V_{\min}| = |V_o^+| (1 - |\Gamma_L|) = 5 \times 0.322 = 1.6 \text{ Volt}$$

$$\text{Max @ } \phi - 2\beta l = 0 \rightarrow \beta l = \frac{85.43}{2} = 42.7^\circ$$

$$\text{Min @ } \phi - 2\beta l = \pi \rightarrow \beta l = \frac{85.43 - 180}{2}$$

at $l = \lambda/2 \quad \beta l = 180$

magnitude of voltage repeats every $l = \frac{\lambda}{2}$

$$\Rightarrow |V(z + \lambda/2)| = |V(z)|$$

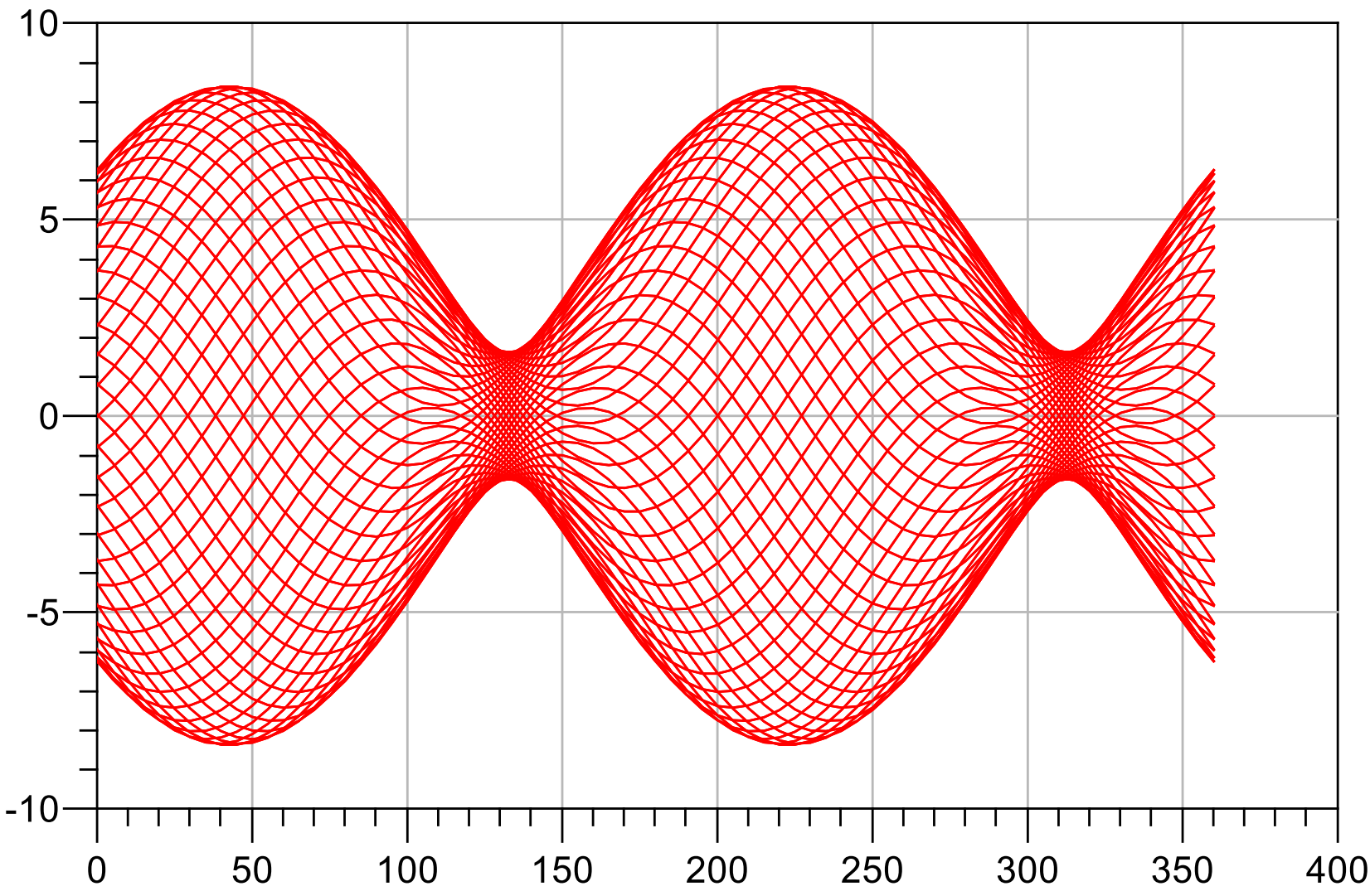
$$\text{or } \left| 1 + |\Gamma_L| e^{j\phi} e^{-2j\beta l} e^{-2j\beta \lambda/2} \right| = \left| 1 + |\Gamma_L| e^{j\phi} e^{-2j\beta l} \right|$$

$$= -47.28 + \boxed{180}$$

$$= 132.7^\circ$$

ADS

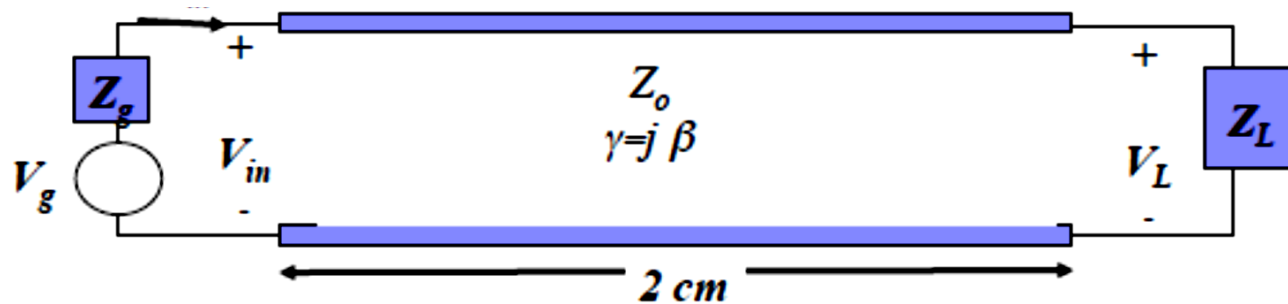
$ts(V1)$



T_length

Example 2

A 2cm lossless TL has $V_g=10$ volt, $Z_g=60 \Omega$, $Z_L=100+j80 \Omega$ and $Z_0=40 \Omega$, $\lambda=10$ cm
Find the input impedance Z_{in} and V_{in}

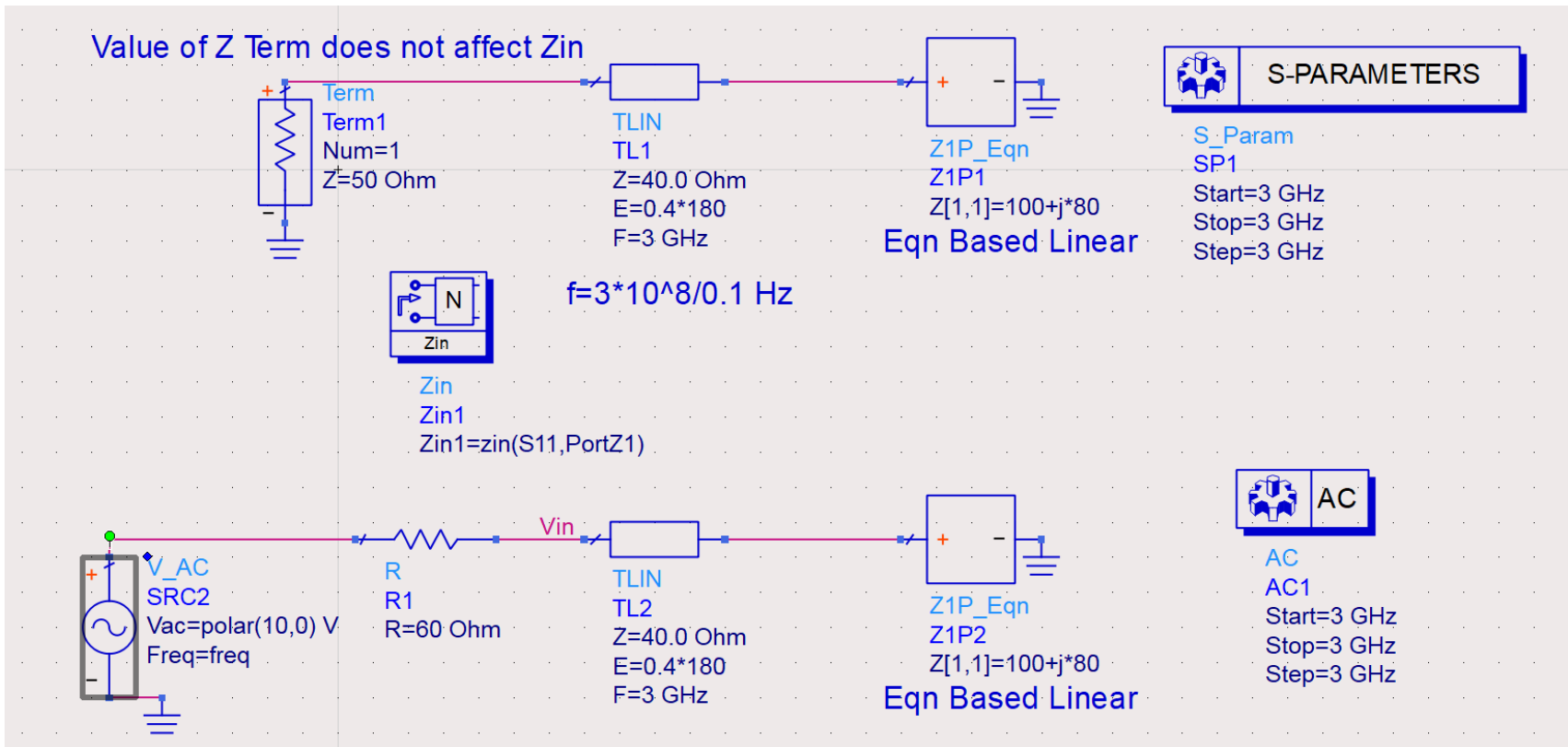


$$Z_{in} = Z_0 \frac{Z_L + jZ_0 \tan \beta l}{Z_0 + jZ_L \tan \beta l} \rightarrow Z_{in} = 12.2 - j21.175$$

$$V_{in} = V_g \frac{Z_{in}}{Z_{in} + Z_g} \rightarrow V_{in} = 2.35 - j2.24$$

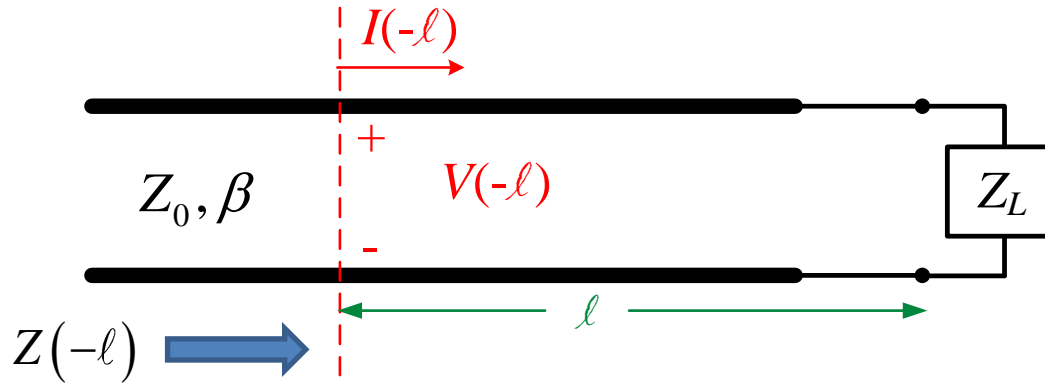
-compute incident voltage at load. (Ans: $3.75\angle -77.75^\circ$)

Example 2 with ADS



freq	Zin1	Vin
3.000 GHz	12.208 - j21.177	2.349 - j2.244

Matched Load



(A) Matched load: ($Z_L = Z_0$)

$$\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0} = 0$$

No reflection from the load

$$\Rightarrow V(-l) = V_0^+ e^{+j\beta l}$$

$$I(-l) = \frac{V_0^+}{Z_0} e^{+j\beta l}$$

$$\Rightarrow Z(-l) = Z_0$$

For any l

Short-Circuit Load

(B) Short circuit load: ($Z_L = 0$)

$$\Gamma_L = \frac{0 - Z_0}{0 + Z_0} = -1$$

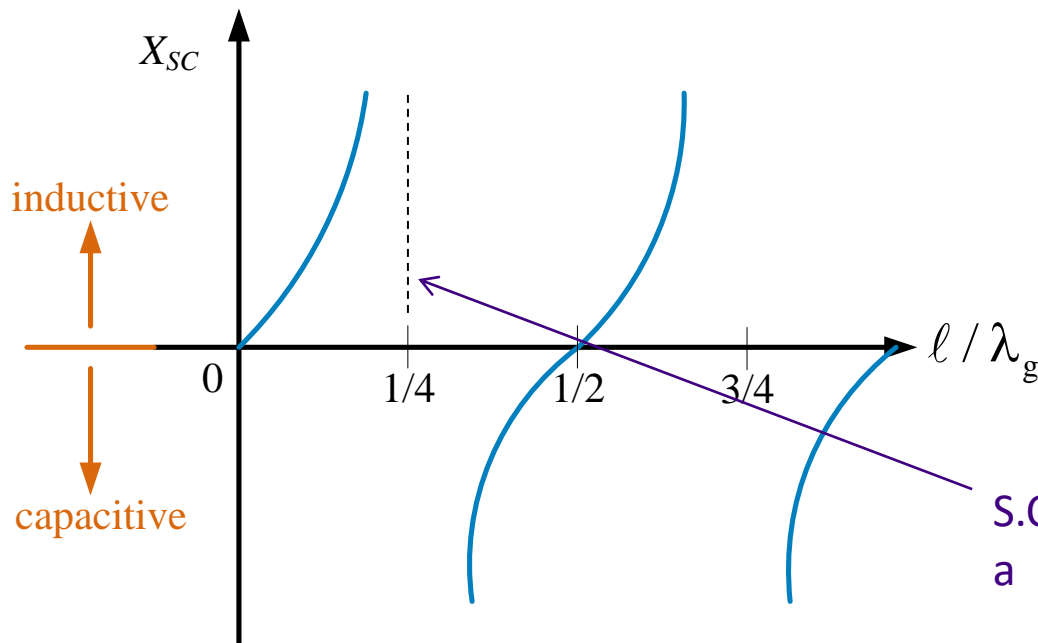
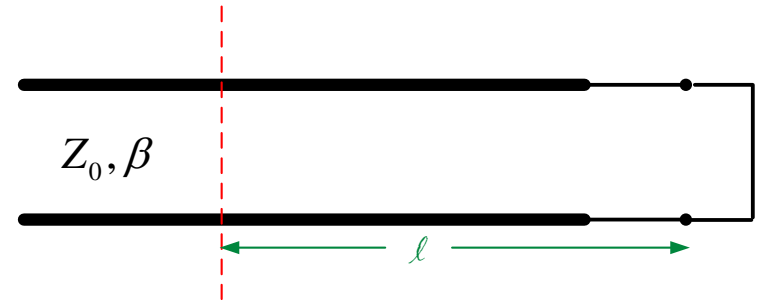
$$\Rightarrow Z(-l) = jZ_0 \tan(\beta l)$$

Note: $\beta l = 2\pi \frac{l}{\lambda_g}$

Always imaginary!

$$\Rightarrow Z(-l) = jX_{sc}$$

$$X_{sc} = Z_0 \tan(\beta l)$$



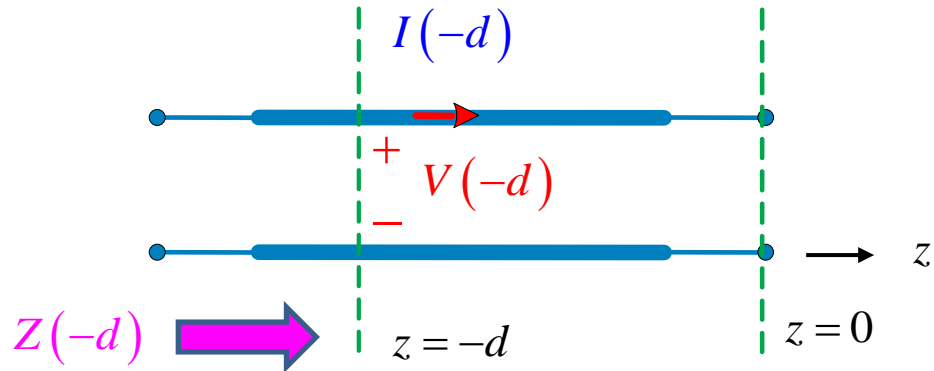
S.C. can become an O.C. with a $\lambda_g/4$ trans. line

Open-Circuit Load ($Z_L = \infty$)

(c) Open circuit load: ($Z_L = \infty$)

$$\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0}$$

$$\Gamma_L = \frac{\infty - Z_0}{\infty + Z_0}$$



$$\Gamma_L = +1$$

$$Z(-d) = Z_0 \left(\frac{Z_L + jZ_0 \tan(\beta d)}{Z_0 + jZ_L \tan(\beta d)} \right) \quad \text{or} \quad Z(-d) = Z_0 \left(\frac{1 + j(Z_0 / Z_L) \tan(\beta d)}{(Z_0 / Z_L) + j \tan(\beta d)} \right)$$

$$Z(-d) = -jZ_0 \cot(\beta d)$$

Always imaginary!

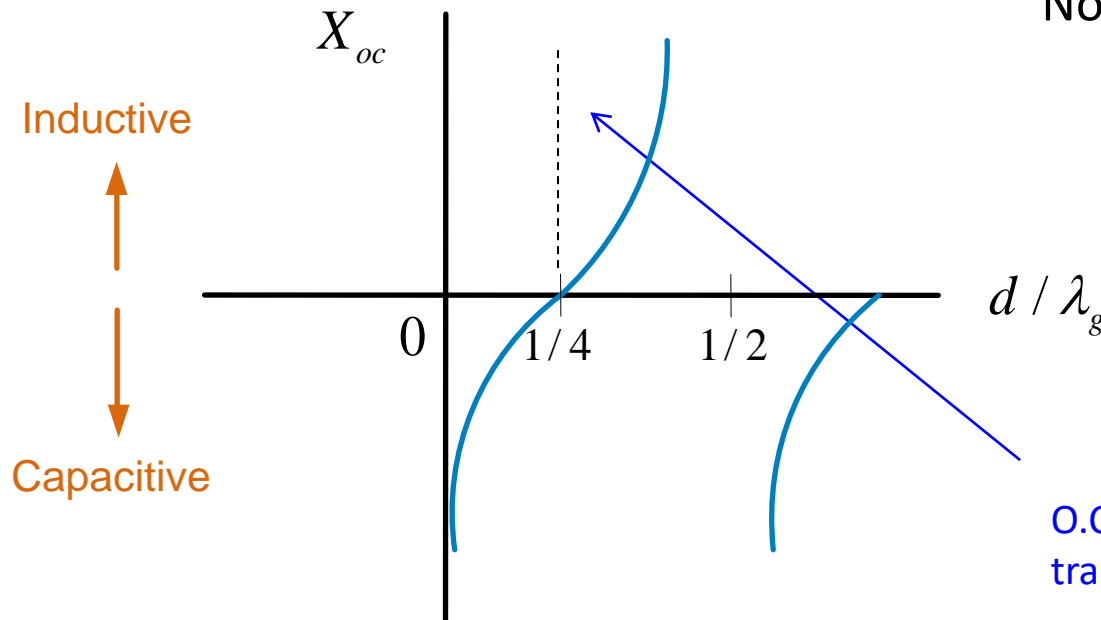
Open-Circuit Load ($Z_L = \infty$)

$$Z(-d) = -jZ_0 \cot(\beta d)$$

$$Z(-d) = jX_{oc}$$

$$X_{oc} = -Z_0 \cot(\beta d)$$

Note: $\beta d = 2\pi \frac{d}{\lambda_g}$



O.C. can become a S.C. with a $\lambda_g/4$ transmission line.

Example 3 Open end and short end TL equivalent elements

S-PARAMETERS

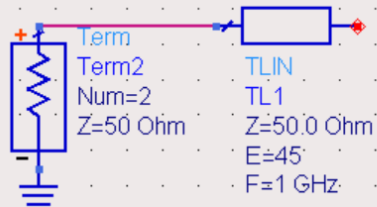
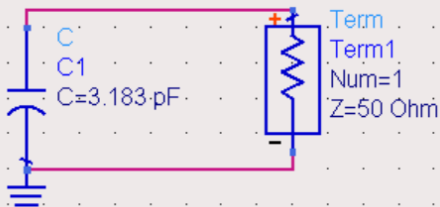
S_Param
SP1
Start=1.0 GHz
Stop=1.0 GHz
Step=1.0 MHz



Zin
Zin1
Zin1=zin(S11,PortZ1)



Zin
Zin2
Zin2=zin(S22,PortZ2)



freq	Zin1	Zin2
1.000 GHz	50.002 / -90.000	50.000 / -90.000

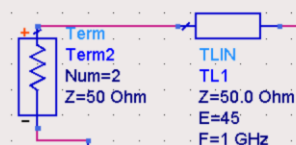
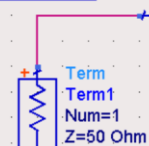
openTLequivC

shortTLequivL

freq	Zin1	Zin2
1.000 GHz	50.014 / 90.000	50.000 / 90.000

S-PARAMETERS

S_Param
SP1
Start=1.0 GHz
Stop=1.0 GHz
Step=1.0 GHz

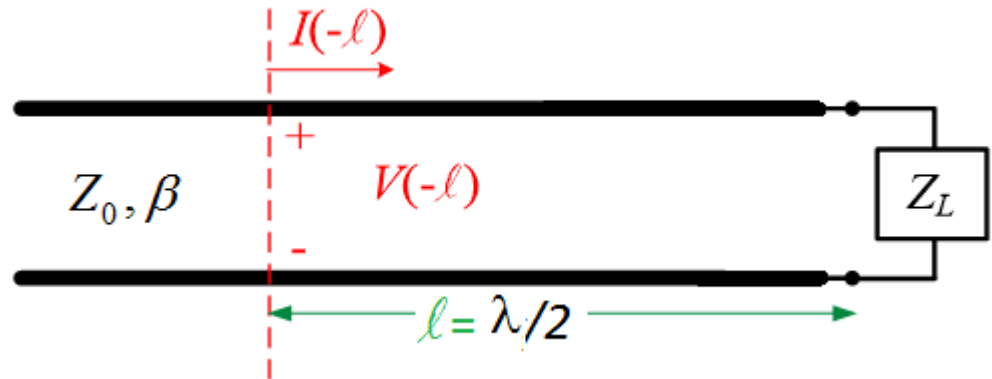


Zin
Zin1
Zin1=zin(S11,PortZ1)



Zin
Zin2
Zin2=zin(S22,PortZ2)

Input impedance for a T.L. of length $l = \lambda/2$

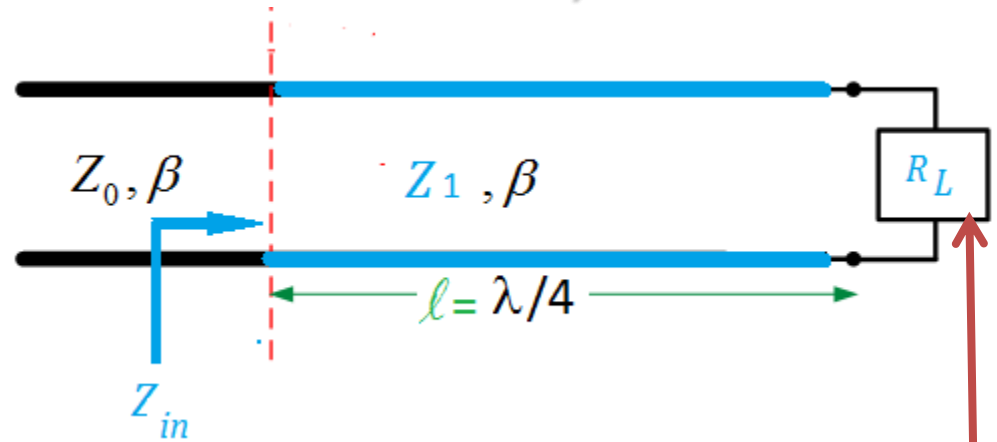


$$Z_{\text{in}} = Z_0 \frac{Z_L + jZ_0 \tan \beta l}{Z_0 + jZ_L \tan \beta l}$$

$$\text{at } l = \lambda/2 \quad \Rightarrow \quad \tan \beta l = 0 \rightarrow Z_{\text{in}} = Z_L$$

So any line(no matter its characteristic impedance) of length $\lambda/2$ or multiple of $\lambda/2$, will look to (or have $Z_{\text{in}} = Z_L$ directly; (as if T.L. does not exist, i.e. T.L. does not transform Z_L at its input)

Input impedance for a T.L. of length $l = \lambda/4$ (quarter wave transformer)



$$Z_{in} = Z_1 \frac{Z_L \cos \beta l + jZ_1 \sin \beta l}{Z_1 \cos \beta l + jZ_L \sin \beta l},$$

$$\text{at } l = \lambda/4, \beta l = \pi/2 \rightarrow Z_{in} = \frac{Z_1^2}{Z_L} = Z_0$$

$$Z_1 = \sqrt{Z_L Z_0} \text{ so input impedance at input of transformer look as } Z_0$$

Example 4

Given a 50Ω transmission line that is 0.25λ long excited by a 1 V voltage source at 300 MHz frequency with an internal impedance of 100Ω , and the line is terminated by a load

$Z_L = 100 - j40 \Omega$, determine $\Gamma_L, Z_{in}, V_{in}, V_o^+$

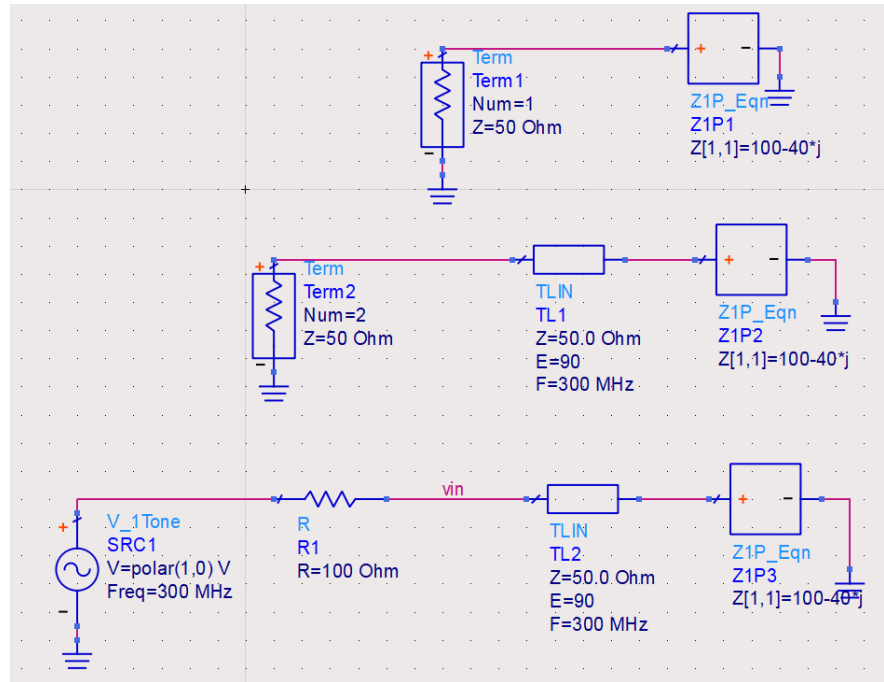
$$\Gamma_L = \frac{Z_L - Z_o}{Z_L + Z_o} = 0.378 - j0.166$$

$$Z_{in} = Z_o * Z_o / Z_L = 21.55 + j8.62$$

$$V_{in} = V_{TH} \frac{Z_{in}}{Z_{in} + Z_{TH}} = 0.1814 + j0.058$$

$$V_o^+ = \frac{V_{in}}{e^{j\beta l} (1 + \Gamma_L e^{-2j\beta l})} = 0.0144 - j0.295$$

Verify by ADS



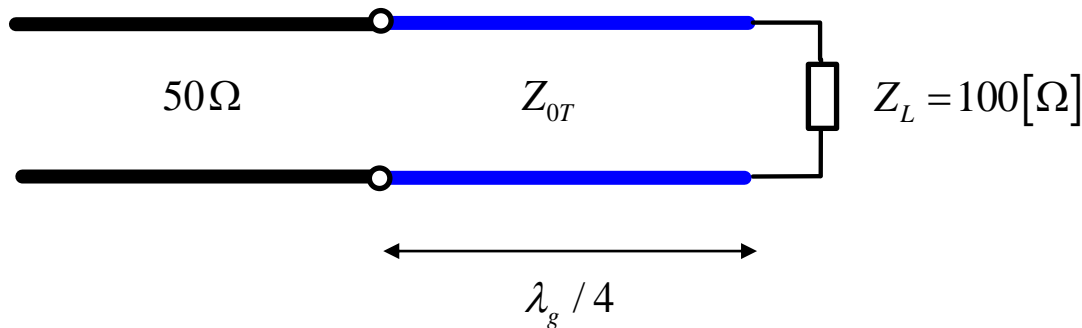
freq	S(1,1)	S(2,2)	Zin1	Zin2	vin
300.0 MHz	0.378 - j0.166	-0.378 + j0.166	107.703 / -21.801	21.552 + j8.621	0.181 + j0.058

Example 5

Match a $100\ \Omega$ load to a $50\ \Omega$ transmission Line at a given frequency.

$$\lambda_g = \frac{2\pi}{\beta} = \frac{2\pi}{k} = \frac{2\pi}{k_0\sqrt{\epsilon_r}} = \frac{\lambda_0}{\sqrt{\epsilon_r}}$$

$$\lambda_0 = \frac{c}{f}$$

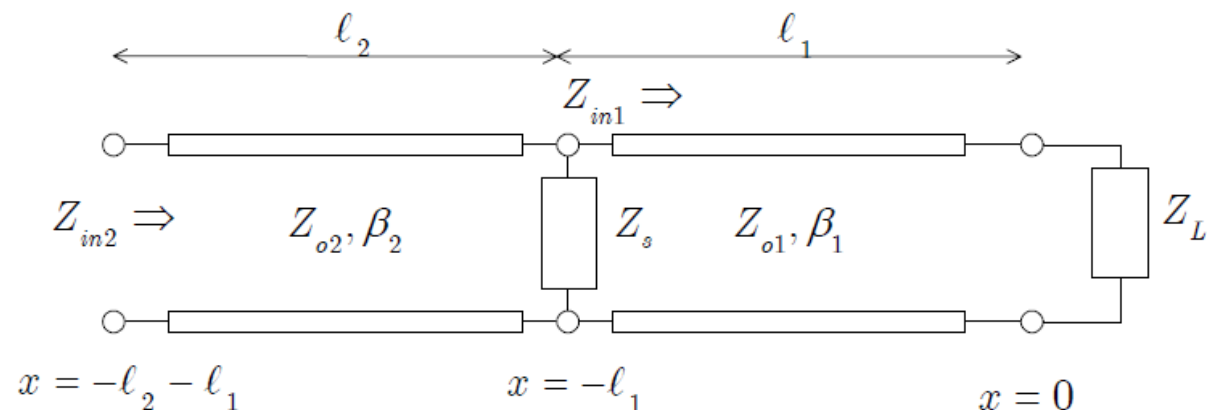


$$\begin{aligned} Z_{0T} &= \sqrt{100 \times 50} \\ &= 70.7 \end{aligned}$$

$$Z_{0T} = 70.7\ \Omega$$

Shunt Loads

A. Parallel Loads



Determine $Z_{in2} = Z_2(-\ell_1 - \ell_2)$:

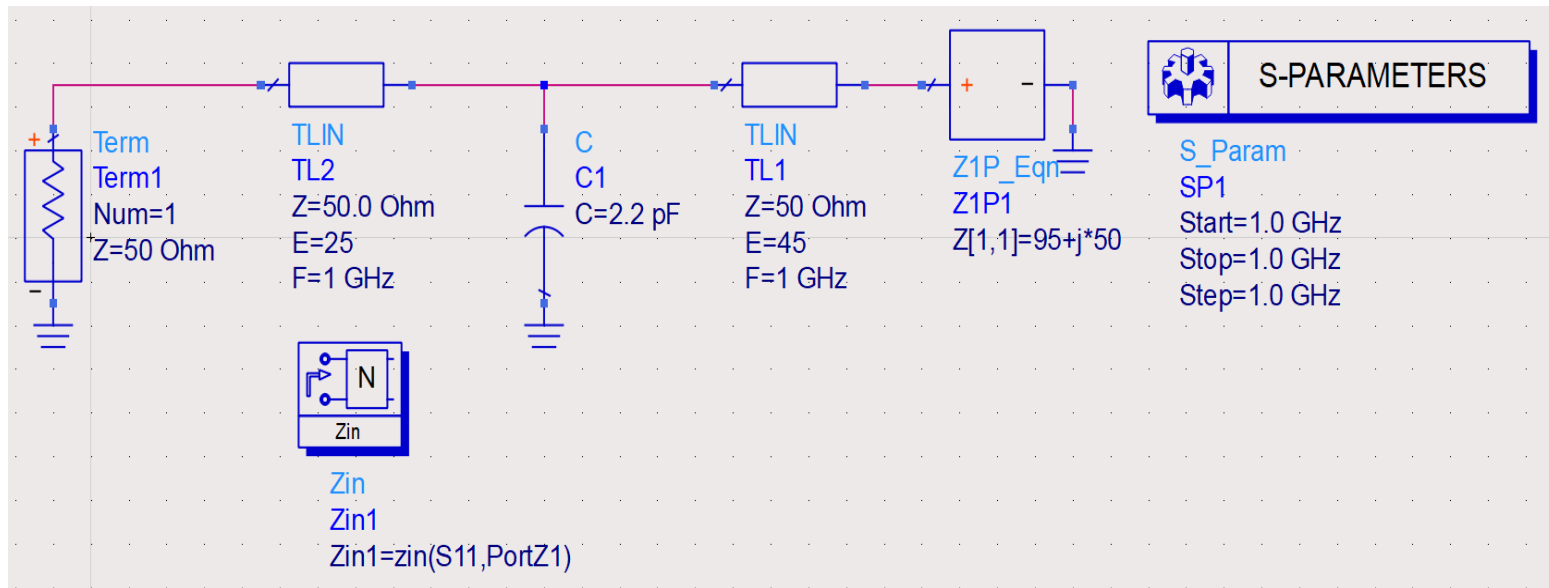
Solution Procedure:

- 1) Apply impedance match at $x=0$
- 2) Determine Z_{in1}
- 3) combine Z_{in1} with Z_s (How do we do this?)
- 4) Determine Z_{in2}

Solution:

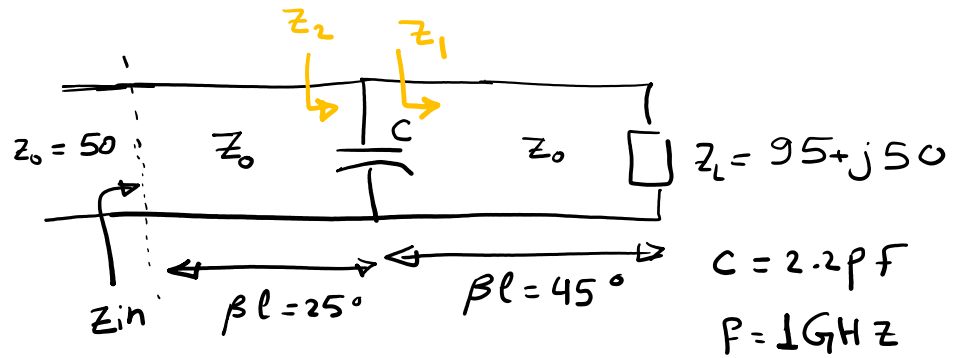
$$Z_{in1} = Z_o \frac{Z_L + jZ_o \tan(\beta\ell_1)}{Z_o + jZ_L \tan(\beta\ell_1)}, \quad Z_{\parallel} = \frac{Z_s Z_{in1}}{Z_s + Z_{in1}}, \quad Z_{in2} = Z_o \frac{Z_{\parallel} + jZ_o \tan(\beta\ell_2)}{Z_o + jZ_{\parallel} \tan(\beta\ell_2)}$$

Example 6



freq	Zin1	S(1,1)
1.000 GHz	10.437 - j10.794	0.668 / -154.614

Solution



$$Z_1 = 50 \frac{(95 + j50) + j50 \tan 45}{50 + j(95 + j50) \tan 45}$$

$$Z_1 = 52.6 - j50 \rightarrow Y_1 = 0.01 + j0.0094$$

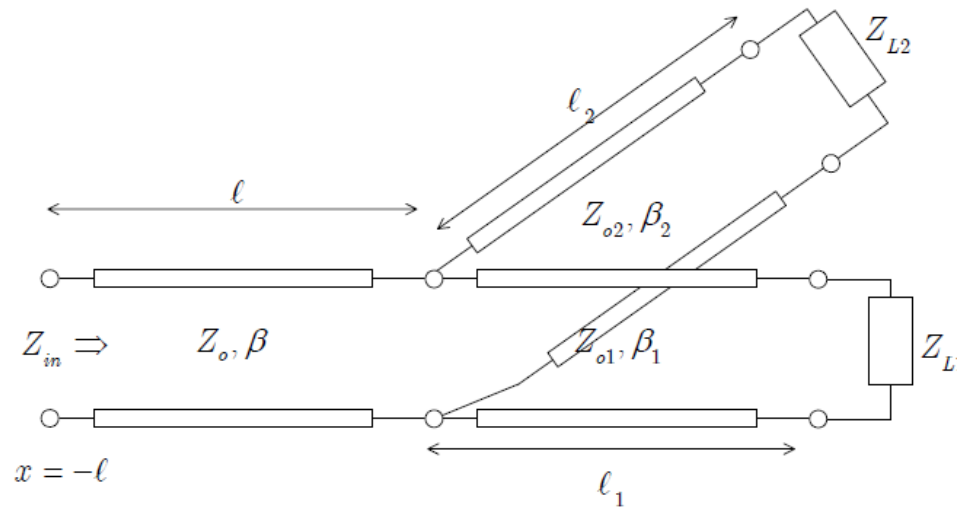
$$j\omega C = j2\pi \times 10^9 \times 2.2 \times 10^{-12} = j0.0138$$

$$Y_2 = Y_1 + j\omega C = 0.01 + j0.023 \Rightarrow Z_2 = 15.9 - j36.6$$

$$Z_{in} = 50 \frac{(15.9 - j36.6) + j50 \tan 25}{50 + j(15.9 - j36.6) \tan 25} = 10.6 - j11 \quad \left[\begin{array}{l} \text{Not Exact} \\ \text{due to Approx.} \\ \text{across solution} \end{array} \right]$$
$$\approx 10.4 - j10.8$$

$$\Gamma_{in} = \frac{Z_{in} - 50}{Z_{in} + 50} = -0.596 - j0.29 = 0.66 \angle -154$$

Parallel Lines:



Determine Z_{in} :

Solution Procedure:

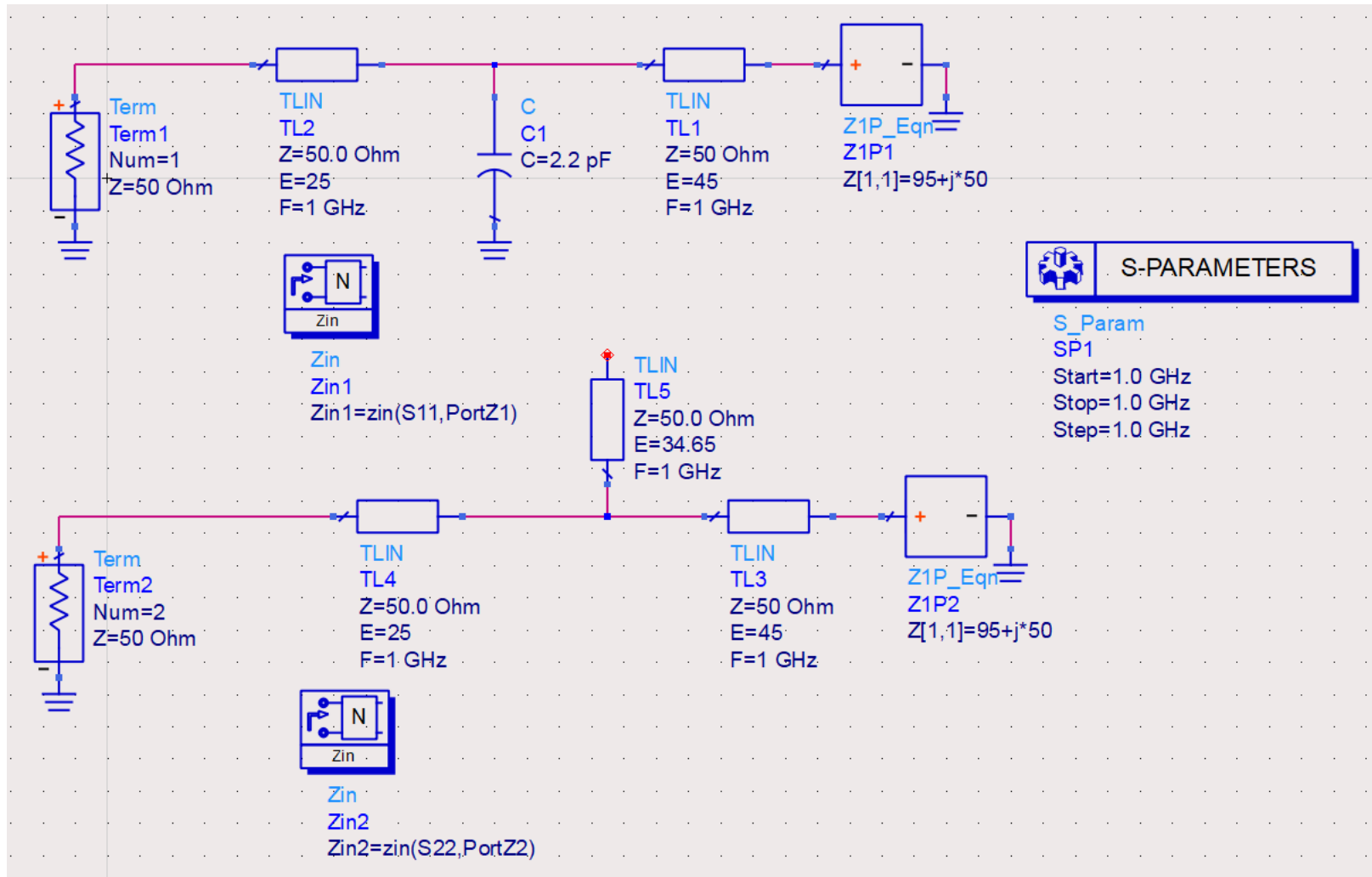
- 1) Determine Z_{in} of lines 1 and 2
- 2) Determine effective load (how do they combine?)
- 3) Determine Z_{in}

Solution:

$$Z_{in1} = Z_{o1} \frac{Z_{L1} + jZ_{o1} \tan(\beta_1 \ell_1)}{Z_{o1} + jZ_{L1} \tan(\beta_1 \ell_1)}, \quad Z_{in2} = Z_{o2} \frac{Z_{L2} + jZ_{o2} \tan(\beta_2 \ell_2)}{Z_{o2} + jZ_{L2} \tan(\beta_2 \ell_2)}$$

$$Z_{\parallel} = \frac{Z_{in1} Z_{in2}}{Z_{in1} + Z_{in2}}, \quad Z_{in} = Z_o \frac{Z_{\parallel} + jZ_o \tan(\beta \ell)}{Z_o + jZ_{\parallel} \tan(\beta \ell)}$$

Example 6 with parallel o.c. TL instead of shunt C



freq	Zin1	S(1,1)	Zin2	S(2,2)
1.000 GHz	10.437 - j10.794	0.668 / -154.614	10.437 - j10.794	0.668 / -154.614

